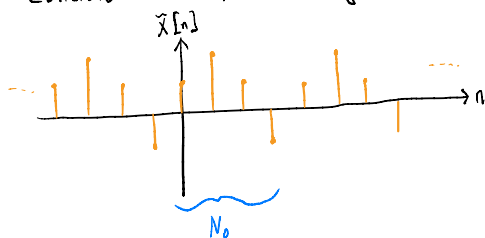


Discrete-Time Fourier Transform

Consider DT periodic signal $\tilde{x}[n]$



Can represent $\tilde{x}[n]$ in frequency domain using DT Fourier series

→ N_0 coefficients a_k

→ $a_0, a_1, \dots, a_{N_0-1}$

$$\tilde{x}[n] = \sum_{k=\langle N_0 \rangle} a_k e^{j2\pi \underbrace{k f_0}_{=f} n} = \sum_{k=\langle N_0 \rangle} a_k e^{j2\pi \left(\frac{k}{N_0}\right) n}$$

$$a_k = \frac{1}{N_0} \sum_{n=\langle N_0 \rangle} \tilde{x}[n] e^{-j2\pi \left(\frac{k}{N_0}\right) n}$$

Like CT FS, each integer k corresponds to a frequency

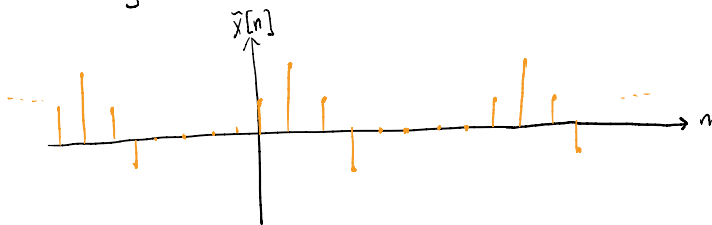
$$\rightarrow f = k f_0 = \frac{k}{N_0}$$

$\tilde{x}[n]$ has a certain "amount" of each frequency $f = k f_0 = \frac{k}{N_0}$ for $k = 0, 1, \dots, N_0-1$

$$\rightarrow f = 0, \frac{1}{N_0}, \frac{2}{N_0}, \dots, \frac{N_0-1}{N_0}$$

→ Each freq. component has a mag/phase set by a_k

Let's try "padding" the periodic signal with zeros

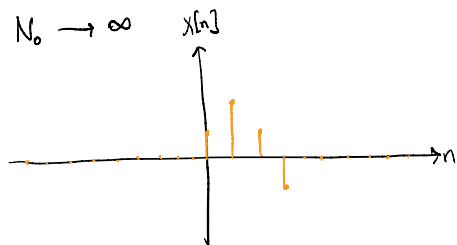


→ Periodic, but increased N_0

→ $f = \frac{k}{N_0}$ gets more closely spaced

Last frequency $f = \frac{N_0-1}{N_0} \rightarrow 1$

Let $N_0 \rightarrow \infty$



$\tilde{x}[n]$ above is periodic

→ Signal no longer periodic → $x[n]$

→ Freq. spacing $k/N_0 \rightarrow 0$

→ f becomes a continuous variable $0 \leq f < 1$

So w/ $N_0 = \infty$, we have $a_k = \frac{1}{N_0} \sum_{n=-\infty}^{\infty} x[n] e^{-j2\pi(\frac{k}{N_0})n}$

Define: $X(f) = \sum_{n=-\infty}^{\infty} x[n] e^{-j2\pi fn}$

Then DT FS coefficients a_k given by $a_k = \frac{1}{N_0} X(f) \Big|_{f=\frac{k}{N_0}=kf_0} = f_0 X(f) \Big|_{f=kf_0}$

Plg this into reconstruction of periodic $\tilde{x}[n]$:

$$\tilde{x}[n] = \sum_{k \in \mathbb{Z}} a_k e^{j2\pi kf_0 n} = \sum_{k \in \mathbb{Z}} X(f = kf_0) e^{j2\pi kf_0 n} \cdot f_0$$

As $N_0 \rightarrow \infty$, $f_0 \rightarrow 0$, final frequency $f = \frac{N_0^{-1}}{N_0} \rightarrow 1$, and summation $\rightarrow$ integral

$\rightarrow \tilde{x}[n] \rightarrow x[n] = \int_0^1 X(f) e^{j2\pi fn} df$

Periodic Aperiodic

$\rightarrow$ Integral for reconstruction of aperiodic signal from continuous spectrum of freq. components!

Discrete-Time Fourier Transform (DTFT) Pair

$$x[n] = \int_{\langle 1 \rangle} X(f) e^{j2\pi fn} df = \frac{1}{2\pi} \int_{\langle 2\pi \rangle} X(\omega) e^{j\omega n} d\omega$$

$$X(f) = \sum_{n=-\infty}^{\infty} x[n] e^{-j2\pi fn} \quad \text{or} \quad X(\omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

From DT FS: a_k is periodic because frequencies are not unique

$\rightarrow$ Same concept in DTFT!

$X(f)$ is periodic w/ period of 1 $\rightarrow 0 \leq f < 1$ OR $-\frac{1}{2} \leq f < \frac{1}{2}$

$X(\omega)$ is periodic w/ period of 2π $\rightarrow 0 \leq \omega < 2\pi$ OR $-\pi \leq \omega < \pi$

Some DTFT Properties

1.) Signal energy

Aperiodic signal $x[n] \rightarrow$ energy signal

Parseval's Theorem (again):

$$E_x = \sum_{n=-\infty}^{\infty} |x[n]|^2 = \int_{(-1)}^{(1)} |X(f)|^2 df = \frac{1}{2\pi} \int_{(-2\pi)}^{(2\pi)} |X(\omega)|^2 d\omega$$

2.) Symmetry: Same as other Fourier representations!

- $x[n]$ real-valued $\rightarrow$ conjugate symmetry, $X(f) = X^*(-f)$
- $x[n]$ real & even $\rightarrow X(f)$ real-valued & even, $X(f) = X(-f)$
- $x[n]$ real & odd $\rightarrow X(f)$ imaginary-valued & odd, $X(f) = -X(-f)$

3.) Discreteness/Periodicity Duality

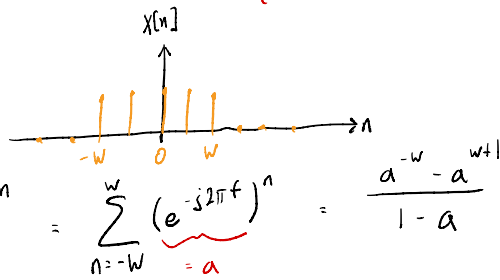
Signals that are **discrete** in one domain are **periodic** in the other

	<u>Time-Domain</u>	<u>Frequency Domain</u>
CT Fourier Transform:	Continuous, Aperiodic	Aperiodic, Continuous
CT Fourier Series:	Continuous, Periodic	Aperiodic, Discrete
DT Fourier Transform:	Discrete, Aperiodic	Periodic, Continuous
DT Fourier Series:	Discrete, Periodic	Periodic, Discrete

Example:

$$x[n] = \begin{cases} 1 & |n| \leq W \\ 0 & \text{else} \end{cases}$$

(Sort of like a DT rect function)



$$X(f) = \sum_{n=-\infty}^{\infty} x[n] e^{-j2\pi f n} = \sum_{n=-W}^W (1) e^{-j2\pi f n} = \sum_{n=-W}^W \underbrace{(e^{-j2\pi f})^n}_{=a} = \frac{a^{-W} - a^{W+1}}{1-a}$$

$$= \frac{e^{j2\pi f W} - e^{-j2\pi f (W+1)}}{1 - e^{-j2\pi f}}$$

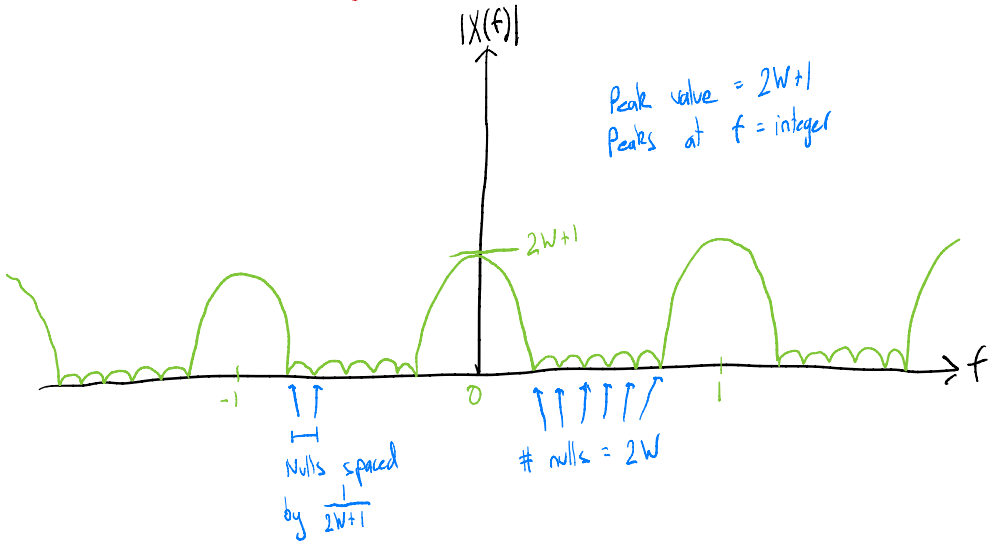
$$= \frac{e^{-j\pi f} e^{j2\pi f (W+\frac{1}{2})} - e^{-j\pi f} e^{-j2\pi f (W+\frac{1}{2})}}{e^{-j\pi f} (e^{j\pi f} - e^{-j\pi f})}$$

$$= \frac{2j \sin(2\pi f (W+\frac{1}{2}))}{2j \sin(\pi f)}$$

$$\boxed{\frac{\sin(\pi f (2W+1))}{\sin(\pi f)}}$$

→ Result: sort of like a sinc function (rect/sinc duality)
But this one is periodic!

→ Numerator & denominator go to zero for integer f regardless of W !



Example: $x[n] = \delta[n] \rightarrow X(f) = \sum_{n=0} (1) e^{-j2\pi f n} = \boxed{1}$

$$x[n] = \delta[n-n_0] \rightarrow X(f) = \sum_{n=n_0} (1) e^{-j2\pi f n} = \boxed{e^{-j2\pi f n_0}}$$

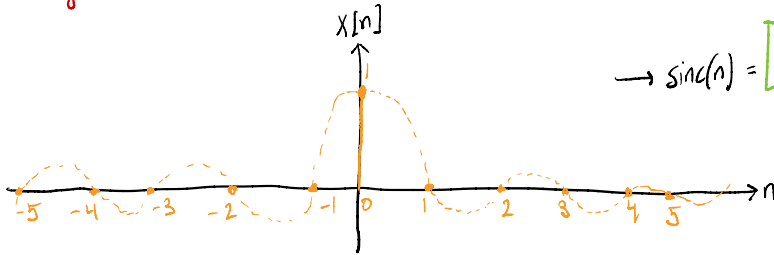
Property: $x[n] \xleftrightarrow{F} X(f) \rightarrow x[n-n_0] \xleftrightarrow{F} X(f) e^{-j2\pi f n_0}$

Inverse DTFT Example: $X(f) = 1$

$$\rightarrow x[n] = \int_{-1/2}^{1/2} X(f) e^{j2\pi f n} df = \int_{-1/2}^{1/2} e^{j2\pi f n} df = \left. \frac{e^{j2\pi f n}}{j2\pi n} \right|_{-1/2}^{1/2}$$

$$= \frac{e^{j\pi n} - e^{-j\pi n}}{j2\pi n} = \frac{j2 \sin(\pi n)}{j2\pi n} = \frac{\sin(\pi n)}{\pi n} = \boxed{\text{sinc}(n)}$$

→ For integer n , $\text{sinc}(n) = 1$ at $n=0$ and $\text{sinc}(n) = 0$ for other n

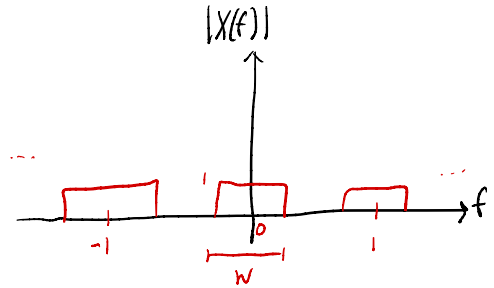
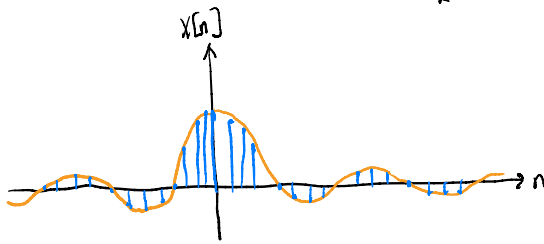


$$\rightarrow \text{sinc}(n) = \boxed{\delta[n] = x[n]}$$

Other important DTFT pairs (on your table!):

$$x[n] = 1 \quad \xleftrightarrow{\mathcal{F}} \quad X(f) = \sum_{k=-\infty}^{\infty} \delta(f-k)$$

$$x[n] = W \text{sinc}(Wn) \quad \xleftrightarrow{\mathcal{F}} \quad X(f) = \sum_{k=-\infty}^{\infty} \text{rect}\left(\frac{f-k}{W}\right)$$



$$x[n] = \cos(2\pi f_0 n) \quad \xleftrightarrow{\mathcal{F}} \quad X(f) = \sum_{k=-\infty}^{\infty} \frac{1}{2} \left(\delta(f-f_0-k) + \delta(f+f_0-k) \right)$$

$$x[n] = a^n u[n] \quad (|a| < 1) \quad \xleftrightarrow{\mathcal{F}} \quad X(f) = \frac{1}{1 - a e^{-j2\pi f}}$$

Other Important Properties (on table):

Frequency Shifting: $x[n] e^{j2\pi f_0 n} \xleftrightarrow{\mathcal{F}} X(f - f_0)$

Convolution: $x[n] * h[n] \xleftrightarrow{\mathcal{F}} X(f) H(f)$ → Just like CT! $h[n]$ impulse response
→ $H(f)$ frequency response for DT systems

Multiplication: $x[n] y[n] \xleftrightarrow{\mathcal{F}} X(f) * Y(f)$ (periodic convolution... we'll probably try to avoid this property)

★ Pairs and properties table for DTFT

→ Use just like CT FT to evaluate complex DTFTs without direct evaluation
→ Note: DTFT of short, finite sequences is easy to compute directly from the summation definition

Example: $x[n] = 2(0.9)^n \cos\left(\frac{\pi}{4}n\right) u[n]$ $\rightarrow$ Find DTFT $X(f)$

$\rightarrow$ Like CT, break into chunks and use pairs & properties

$\rightarrow$ For these, unlike CT, we want to avoid freq. convolution, but can apply frequency shifting instead

$$\rightarrow \cos\left(\frac{\pi}{4}n\right) = \frac{1}{2}e^{j\frac{\pi}{4}n} + \frac{1}{2}e^{-j\frac{\pi}{4}n}$$

$$\rightarrow x[n] = (0.9)^n e^{j\frac{\pi}{4}n} + (0.9)^n e^{-j\frac{\pi}{4}n}$$

Pair: $a^n u[n] \xleftrightarrow{\mathcal{F}} \frac{1}{1 - ae^{-j2\pi f}}$

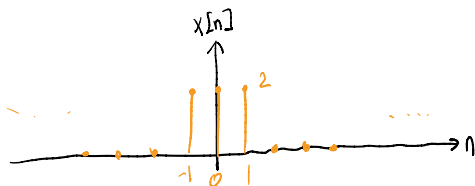
$$\rightarrow (0.9)^n u[n] \xleftrightarrow{\mathcal{F}} \frac{1}{1 - 0.9e^{-j2\pi f}}$$

Frequency-Shifting Property: $x[n] e^{j2\pi f_0 n} \xleftrightarrow{\mathcal{F}} X(f - f_0)$

For $e^{\pm j\frac{\pi}{4}n} \rightarrow f_0 = \pm \frac{1}{8}$

$$\rightarrow X(f) = \frac{1}{1 - 0.9e^{j2\pi(f - \frac{1}{8})}} + \frac{1}{1 - 0.9e^{j2\pi(f + \frac{1}{8})}}$$

Another Example:



a.) Find DTFT of $\tilde{x}[n] = 2x[n] \cos(2\pi(\frac{1}{3})n)$

First: $x[n] = \begin{cases} 2 & |n| \leq 1 \\ 0 & \text{else} \end{cases} \xleftrightarrow{F} X(f) = \frac{\sin(\pi f (2W+1))}{\sin(\pi f)} = 2 \frac{\sin(3\pi f)}{\sin(\pi f)}$

Euler's: $\tilde{x}[n] = 2x[n] \left(\frac{1}{2} e^{j2\pi(\frac{1}{3})n} + \frac{1}{2} e^{-j2\pi(\frac{1}{3})n} \right) = x[n] e^{j2\pi(\frac{1}{3})n} + x[n] e^{-j2\pi(\frac{1}{3})n}$

→ Frequency shift by $f_0 = \pm \frac{1}{3}$

→ $\tilde{X}(f) = X(f - \frac{1}{3}) + X(f + \frac{1}{3}) = 2 \frac{\sin(3\pi(f - \frac{1}{3}))}{\sin(\pi(f - \frac{1}{3}))} + 2 \frac{\sin(3\pi(f + \frac{1}{3}))}{\sin(\pi(f + \frac{1}{3}))}$

b.) Let $h[n] = \delta[n-10]$ Find $H(f)$.

Pair: $\delta[n-n_0] \xleftrightarrow{F} e^{-j2\pi f n_0}$

→ $H(f) = e^{-j20\pi f}$

c.) Let $y[n] = \tilde{x}[n] * h[n]$. Find $Y(f)$ and take the Inverse DTFT to find $y[n]$.

Convolution: $Y(f) = \tilde{X}(f) H(f) = 2 \left(\frac{\sin(3\pi(f - \frac{1}{3}))}{\sin(\pi(f - \frac{1}{3}))} + \frac{\sin(3\pi(f + \frac{1}{3}))}{\sin(\pi(f + \frac{1}{3}))} \right) e^{-j20\pi f}$

$y[n]?$ → $2 \frac{\sin(3\pi(f - f_0))}{\sin(\pi(f - f_0))} \xleftrightarrow{F^{-1}} x[n] e^{j2\pi f_0 n}$

Phase shift in freq.
→ Time shift

$2 \frac{\sin(3\pi(f - f_0))}{\sin(\pi(f - f_0))} e^{-j2\pi f n_0} \xleftrightarrow{F^{-1}} x[n - n_0] e^{j2\pi f(n - n_0)}$

→ $y[n] = x[n-10] e^{j2\pi(\frac{1}{3})(n-10)} + x[n-10] e^{-j2\pi(\frac{1}{3})(n-10)} = 2x[n-10] \cos(2\pi(\frac{1}{3})(n-10))$

$= \tilde{x}[n-10] = \tilde{x}[n] * \delta[n-10]$